

STD60NH03L

N-CHANNEL 30V - $0.0075~\Omega$ - 60A DPAK/IPAK STripFETTM III POWER MOSFET

TYPE	V _{DSS}	R _{DS(on)}	I _D	
STD60NH03L	30 V	< 0.009 Ω	60 A	

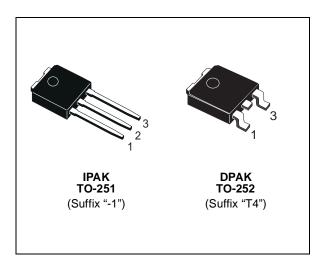
- TYPICAL $R_{DS}(on) = 0.0075 \Omega @ 10 V$
- TYPICAL $R_{DS}(on) = 0.009 \Omega @ 5 V$
- R_{DS(ON)} * Qg INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- THROUGH-HOLE IPAK (TO-251) POWER PACKAGE IN TUBE (SUFFIX "-1")
- SURFACE-MOUNTING DPAK (TO-252)
 POWER PACKAGE IN TAPE & REEL (SUFFIX "T4")

DESCRIPTION

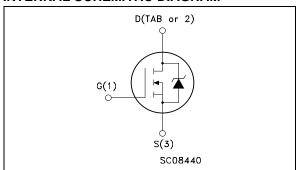
The STD60NH03L utilizes the latest advanced design rules of ST's proprietary STripFET™ technology. This is suitable fot the most demanding DC-DC converter application where high efficiency is to be achieved.

APPLICATIONS

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTES



INTERNAL SCHEMATIC DIAGRAM



Ordering Information

SALES TYPE	MARKING	PACKAGE	PACKAGING
STD60NH03LT4	D60NH03L	TO-252	TAPE & REEL
STD60NH03L-1	D60NH03L	TO-251	TUBE

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{DS}	Drain-source Voltage (V _{GS} = 0)	30	V
V_{DGR}	Drain-gate Voltage ($R_{GS} = 20 \text{ k}\Omega$)	30	V
V_{GS}	Gate- source Voltage	± 20	V
I _D	Drain Current (continuous) at T _C = 25°C	60	A
I _D	Drain Current (continuous) at T _C = 100°C	43	A
I _{DM} (1)	Drain Current (pulsed)	240	A
P _{tot}	Total Dissipation at T _C = 25°C	70	W
	Derating Factor	0.47	W/°C
E _{AS} (2)	Single Pulse Avalanche Energy	300	mJ
T _{stg}	Storage Temperature	-55 to 175	°C
Tj	Max. Operating Junction Temperature	-55 to 175	

October 2003 1/12

THERMAL DATA

Rthj-case	Thermal Resistance Junction-case	Max	2.14	°C/W
Rthj-amb	Thermal Resistance Junction-ambient	Max	100	°C/W
Rthj-pcb	Thermal Resistance Junction-pcb(#)	Max	43	°C/W
Tı	Maximum Lead Temperature For Soldering Purpose		275	°C

^(#) When Mounted on 1 inch² FR-4 board, 2 oz of Cu.

ELECTRICAL CHARACTERISTICS ($T_{CASE} = 25~^{\circ}C$ UNLESS OTHERWISE SPECIFIED) OFF

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
V _{(BR)DSS}	Drain-source Breakdown Voltage	$I_D = 250 \mu A, V_{GS} = 0$	30			V
I _{DSS}	Zero Gate Voltage Drain Current (V _{GS} = 0)	V_{DS} = Max Rating V_{DS} = Max Rating T_{C} = 125°C			1 10	μA μA
I _{GSS}	Gate-body Leakage Current (V _{DS} = 0)	V _{GS} = ± 20V			±100	nA

ON (4)

Symbol	Parameter Test Conditions		Min.	Тур.	Max.	Unit	
V _{GS(th)}	Gate Threshold Voltage	$V_{DS} = V_{GS}$	I _D = 250 μA	1			V
R _{DS(on)}	Static Drain-source On Resistance	V _{GS} = 10 V V _{GS} = 5 V	I _D = 30 A I _D = 30 A		0.0075 0.009	0.009 0.017	Ω Ω

DYNAMIC

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
9fs ⁽⁴⁾	Forward Transconductance	$V_{DS} = 15 \text{ V}$ $I_{D} = 18 \text{ A}$		25		S
C _{iss} C _{oss} C _{rss}	Input Capacitance Output Capacitance Reverse Transfer Capacitance	V _{DS} = 10V f = 1 MHz V _{GS} = 0		2200 380 49		pF pF pF
R _G	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.5		Ω

ELECTRICAL CHARACTERISTICS (continued)

SWITCHING ON

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t _{d(on)} t _r	Turn-on Delay Time Rise Time	$\begin{aligned} V_{DD} &= 15 \text{ V} & I_D &= 30 \text{ A} \\ R_G &= 4.7 \ \Omega & V_{GS} &= 5 \text{ V} \\ \text{(Resistive Load, Figure 3)} \end{aligned}$		21 95		ns ns
Q _g Q _{gs} Q _{gd}	Total Gate Charge Gate-Source Charge Gate-Drain Charge	V _{DD} = 15 V I _D = 60 A V _{GS} = 5 V		15.7 8.3 3.4	21	nC nC nC
Q _{gls} (4)	Third-quadrant Gate Charge	V _{DS} < 0 V V _{GS} = 10 V		15		nC

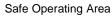
SWITCHING OFF

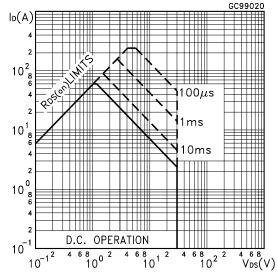
I	Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
	$t_{d(off)}$ t_{f}	Turn-off Delay Time Fall Time	$\begin{array}{ccc} V_{DD} = 15 \text{ V} & I_D = 30 \text{ A} \\ R_G = 4.7\Omega, & V_{GS} = 5 \text{ V} \\ \text{(Resistive Load, Figure 3)} \end{array}$		19 15		ns ns

SOURCE DRAIN DIODE

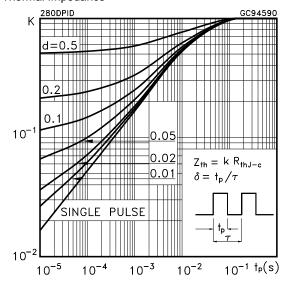
Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
I _{SD} I _{SDM}	Source-drain Current Source-drain Current (pulsed)					60 240	A A
V _{SD}	Forward On Voltage	I _{SD} = 30 A	$V_{GS} = 0$			1.4	V
t _{rr} Q _{rr} I _{RRM}	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	I _{SD} = 60 A V _{DD} = 20 V (see test circu	di/dt = $100A/\mu s$ $T_j = 150$ °C it, Figure 5)		32 51 3.2		ns nC A

⁽¹⁾ Pulse width limited by safe operating area (2) Starting $T_j = 25$ °C, $I_D = 30$ A, $V_{DD} = 20$ V



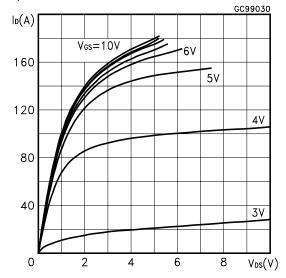


Thermal Impedance

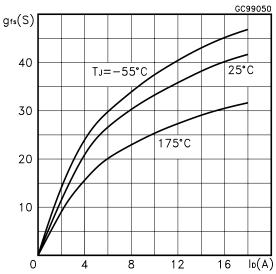


⁽³⁾ Pulsed: Pulse duration = 300 µs, duty cycle 1.5 %. (4) Gate charge for synchronous operation. See Appendix A

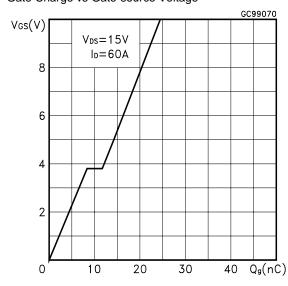
Output Characteristics



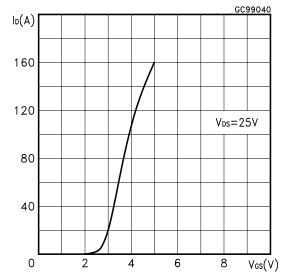
Transconductance



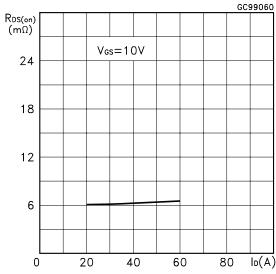
Gate Charge vs Gate-source Voltage



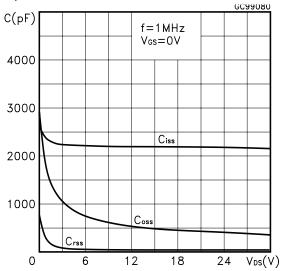
Transfer Characteristics



Static Drain-source On Resistance



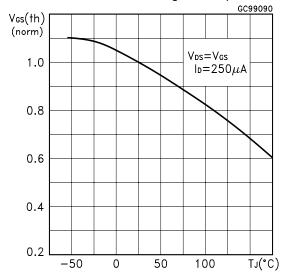
Capacitance Variations



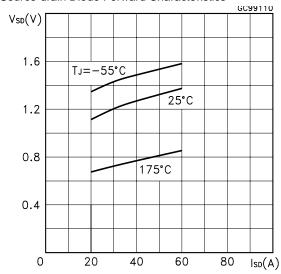
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STD60NH03L

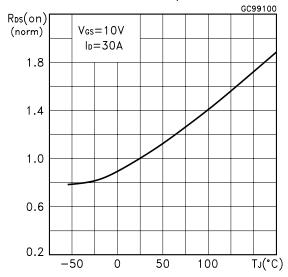
Normalized Gate Threshold Voltage vs Temperature



Source-drain Diode Forward Characteristics



Normalized on Resistance vs Temperature



Normalized Breakdown Voltage vs Temperature

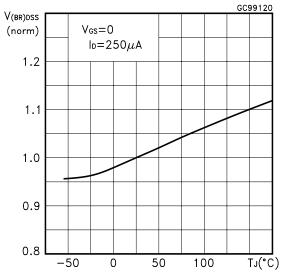


Fig. 1: Unclamped Inductive Load Test Circuit

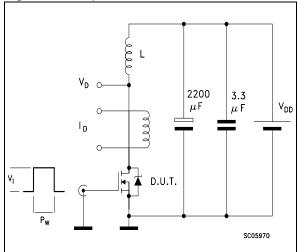


Fig. 3: Switching Times Test Circuits For Resistive Load

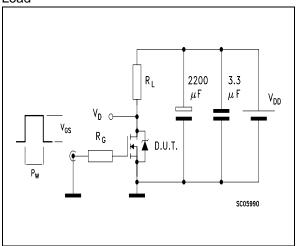


Fig. 5: Test Circuit For Inductive Load Switching And Diode Recovery Times

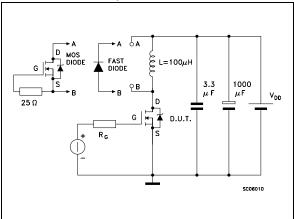


Fig. 2: Unclamped Inductive Waveform

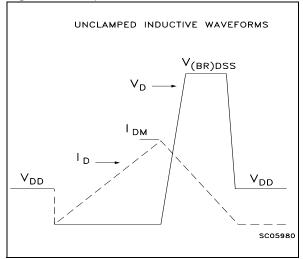
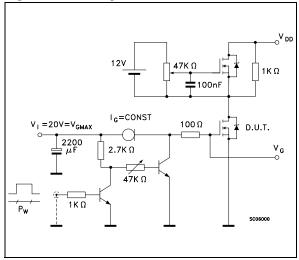
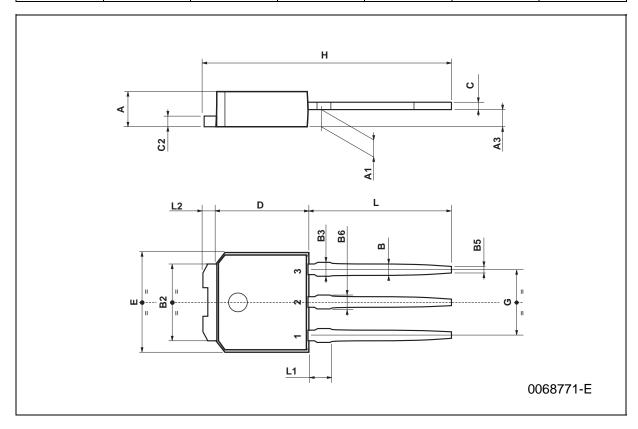


Fig. 4: Gate Charge test Circuit



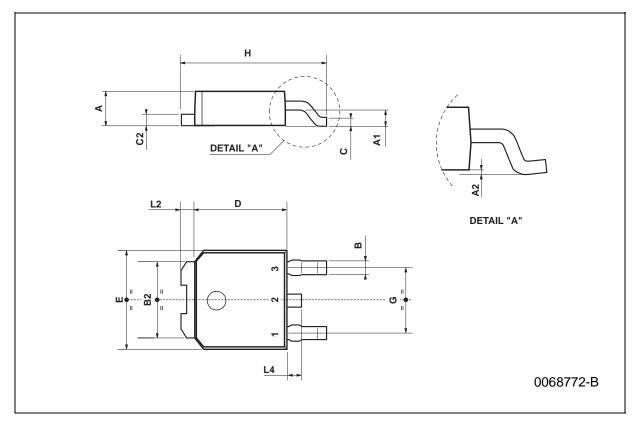
TO-251 (IPAK) MECHANICAL DATA

DIM.		mm			inch	
DIIVI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
А3	0.7		1.3	0.027		0.051
В	0.64		0.9	0.025		0.031
B2	5.2		5.4	0.204		0.212
В3			0.85			0.033
B5		0.3			0.012	
B6			0.95			0.037
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
E	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	15.9		16.3	0.626		0.641
L	9		9.4	0.354		0.370
L1	0.8		1.2	0.031		0.047
L2		0.8	1		0.031	0.039



TO-252 (DPAK) MECHANICAL DATA

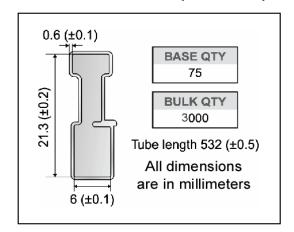
DIM.		mm			inch	
Diwi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A2	0.03		0.23	0.001		0.009
В	0.64		0.9	0.025		0.035
B2	5.2		5.4	0.204		0.212
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
Е	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	9.35		10.1	0.368		0.397
L2		0.8			0.031	
L4	0.6		1	0.023		0.039



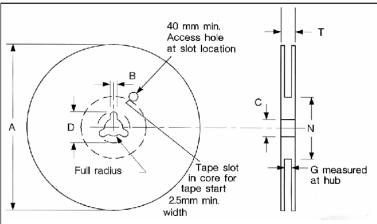
DPAK FOOTPRINT

6.7 1.8 3.0 1.6 2.3 1.6 All dimensions are in millimeters

TUBE SHIPMENT (no suffix)*



TAPE AND REEL SHIPMENT (suffix "T4")*



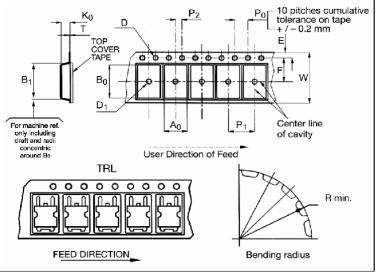
REEL MECHANICAL DATA

DIM.	mm		inch	
	MIN.	MAX.	MIN.	MAX.
Α		330		12.992
В	1.5		0.059	
С	12.8	13.2	0.504	0.520
D	20.2		0.795	
G	16.4	18.4	0.645	0.724
Ν	50		1.968	
Т		22.4		0.881

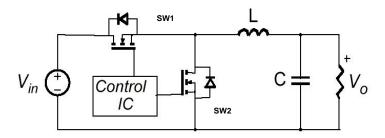
TAPE MECHANICAL	DATA

DIM.	mm		inch	
Dilvi.	MIN.	MAX.	MIN.	MAX.
A0	6.8	7	0.267	0.275
В0	10.4	10.6	0.409	0.417
B1		12.1		0.476
D	1.5	1.6	0.059	0.063
D1	1.5		0.059	
Е	1.65	1.85	0.065	0.073
F	7.4	7.6	0.291	0.299
K0	2.55	2.75	0.100	0.108
P0	3.9	4.1	0.153	0.161
P1	7.9	8.1	0.311	0.319
P2	1.9	2.1	0.075	0.082
R	40		1.574	
W	15.7	16.3	0.618	0.641

BASE QTY	BULK QTY
1000	1000



APPENDIX A Buck Converter: Power Losses Estimation



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is removed to allow for a safer working junction temperature.

The low side (SW2) device requires:

- Very low R_{DS(on)} to reduce conduction losses
- ullet Small Q_{gls} to reduce the gate charge losses
- Small C_{oss} to reduce losses due to output capacitance
- Small Q_{rr} to reduce losses on SW₁ during its turn-on
- The C_{gd}/C_{gs} ratio lower than V_{th}/V_{gg} ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (SW1) device requires:

- \bullet $\,$ Small R_g and L_s to allow higher gate current peak and to limit the voltage feedback on the gate
- Small Q_g to have a faster commutation and to reduce gate charge losses
- Low R_{DS(on)} to reduce the conduction losses.

		High Side Switch (SW1)	Low Side Switch (SW2)
Pconducti	ion	$R_{DS(on)SW1} * I_L^2 * d$	$R_{DS(on)SW2} * I_L^2 * (1-d)$
Pswitchin	g	$V_{\text{in}} * (Q_{\text{gsth(SWI)}} + Q_{\text{gd(SWI)}}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P _{diode}	Recovery	Not Applicable	¹ V _{in} *Q _{rr(SW2)} *f
	Conduction	Not Applicable	$V_{\scriptscriptstyle f(SW2)}*I_{\scriptscriptstyle L}*t_{\scriptscriptstyle deadtime}*f$
$P_{\text{gate}(Q_G)}$)	$Q_{g(SW1)}*V_{gg}*f$	$Q_{gls(SW2)}*V_{gg}*f$
P _{Qoss}		$\frac{V_{in} *Q_{oss(SW1)} *f}{2}$	$\frac{V_{in} *Q_{oss(SW2)} *f}{2}$

Parameter	Meaning
d	Duty-cycle
Q_{gsth}	Post threshold gate charge
$Q_{ m gls}$	Third quadrant gate charge
Pconduction	On state losses
Pswitching	On-off transition losses
Pdiode	Conduction and reverse recovery diode losses
Pgate	Gate drive losses
P _{Qoss}	Output capacitance losses

¹ Dissipated by SW1 during turn-on

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